

UHF QUADRATURE MODULATOR

RF2424

Typical Applications

- Digital Communications Systems
- Spread Spectrum Communication Systems
 AM, SSB, DSB Modulation
- GMSK, QPSK, DQPSK, QAM Modulation
- GSM and D-AMPS Systems
- Image-Reject Upconverters

Product Description

The RF2424 is a monolithic integrated guadrature modulator IC capable of universal direct modulation for highfrequency AM, PM, or compound carriers. Maximum output power is +7.5dBm, which is achieved with low input I and Q signal levels. This low-cost IC implements differential amplifiers for the modulation inputs, 90° carrier phase shift network, carrier limiting amplifiers, two matched double-balanced mixers, summing amplifier, and an output RF amplifier which will drive 50Ω from 700MHz to 1000MHz.





Functional Block Diagram



Package Style: SSOP-16

Features

- Single 2.7V to 5.5V Power Supply
- +7.5dBm Output Power
- No Tuning Required
- Low LO Input Level
- Digitally Controlled Power Down Mode
- 700MHz to 1000MHz Operation

Ordering Information RF2424 UHF Quadrature Modulator RF2424 PCBA Fully Assembled Evaluation Board RF Micro Devices, Inc. Tel (336) 664 1233 7628 Thorndike Road Fax (336) 664 0454 Greensboro, NC 27409, USA http://www.rfmd.com 5

Absolute Maximum Ratings

Parameter	Rating	Unit
Supply Voltage	5.5	V _{DC}
Input LO and RF Levels	+6.0	dBm
Operating Ambient Temperature	-40 to +85	°C
Storage Temperature	-40 to +150	°C



RF Micro Devices believes the furnished information is correct and accurate at the time of this printing. However, RF Micro Devices reserves the right to make changes to its products without notice. RF Micro Devices does not assume responsibility for the use of the described product(s).

Parameter	Specification		Unit	Condition		
Farameter	Min.	Тур.	Max.	Unit	Condition	
Carrier Input					T=25 °C, V _{CC} =3V	
Frequency Range	700		1000	MHz		
Power Level	-6		+6	dBm		
Input impedance		43 + j1.8		Ω	At 900MHz	
Modulation Input						
Frequency Range	DC		100	MHz		
Reference Voltage (V _{REF})		1.6		V		
Maximum Modulation (I&Q)		$V_{REF} \pm 0.3$		V		
Gain Asymmetry		0.2		dB		
Quadrature Phase Error		1		0		
Input DC Resistance		40		kΩ		
Input Bias Current		40		μA		
RF Output					V _{CC} =3V, LO power=-3dBm, LO	
Ki Odipat					freq=900MHz, I/Q drive level= $0.2V_P$, SSB	
Output Power	+5.0	+7.5		dBm		
Output Impedance		50		Ω		
Broadband Noise Floor		-140		dBm/Hz		
Sideband Suppression	25	35		dB		
Carrier Suppression	25	30		dB		
IM ₃ Suppression	25	30		dB	DSB output (+9dBm total power)	
Power Down						
Turn On/Off Time			100	ns		
PD Input Resistance	10			kΩ		
Power Down "ON"		V _{CC}		V		
Power Down "OFF"	1.0	1.2		V		
Power Supply						
Voltage	2.7		5.5	V		
Current		45	55	mA	V _{CC} =3V	
		53		mA	V _{CC} =5V	
Power Down			10	μA		

RF2424

Pin	Function	Description	Interface Schematic
2	I REF Q REF	Reference voltage for the I mixer. This voltage should be the same as the DC voltage supplied to the I SIG pin. A voltage of 1.6V is recom- mended. The SIG and REF inputs are inputs of a differential amplifier. Therefore the REF and SIG inputs are interchangeable. If swapping the I SIG and I REF pins, the Q SIG and Q REF also need to be swapped to maintain the correct phase. It is also possible to drive the SIG and REF inputs in a balanced mode. If single ended operation is desired then the input is applied to I SIG, pin 16. In that case, I REF and Q REF are tied together and AC coupled to ground. To obtain a carrier sup- pression of better than 40dB, I REF may be tuned ±20mV relative to the I SIG. Without tuning, it will typically be better than 25dB. Reference voltage for the Q mixer. This voltage should be the same as	Q SIGO BIAS BIAS
		the DC voltage supplied to the Q SIG pin. A voltage of 1.6V is recom- mended. The SIG and REF inputs are inputs of a differential amplifier. Therefore the REF and SIG inputs are interchangeable. If swapping the Q SIG and Q REF pins, the I SIG and I REF also need to be swapped to maintain the correct phase. It is also possible to drive the SIG and REF inputs in a balanced mode. If single ended operation is desired then the input is applied to Q SIG, pin 15. In that case, Q REF and Q REF are tied together and AC coupled to ground. To obtain a carrier suppression of better than 40dB, Q REF may be tuned ±20mV relative to the Q SIG. Without tuning, it will typically be better than 25dB.	BIAS BIAS
3	VCC1	Power supply for the I mixer, Q mixer and the RF Output amplifier.	
4	MIX OUT	MIXOUT: Combined output of the I mixer and Q mixer. By changing the inductor value, maximum RF output is tuned to different frequency. If the inductor value is changed, the RF output match needs to be adjusted for 50Ω output impedance.	
5	GND1	Ground connection for the LO and baseband amplifiers and mixers.	
6	GND1	Same as pin 5.	
7	LO IN	The input of the phase shifting network.	
8	GND3	Ground connection for the LO phase shift network.	
9	RF OUT	RF Output. An external LC matching network is needed for a 50Ω match.	BIAS
10	GND2	Ground connection for the RF output stage.	
11	GND1	Same as pin 5.	
12	PD	Power Down control. When this pin is "low", all circuits are shut off.	PDO
13	VCC2	Power supply for all circuits except mixers and output amplifier.	
14	GND1	Same as pin 5.	
15	Q SIG	Baseband input to the Q mixer. Maximum output power is obtained when the input signal has a peak to peak amplitude of 400mV. The DC level for this pin is 1.6V, same as QREF.	See pin 2.
16	I SIG	Baseband input to the Q mixer. Maximum output power is obtained when the input signal has a peak to peak amplitude of 400mV. The DC level for this pin is 1.6V, same as QREF.	See pin 1.

5

RF2424

Application Schematic



NOTE:

The values of R and C depend on the minimum baseband frequency (i.e., the cutoff frequency of this high pass filter should be lower than the lowest frequency component in the I/Q spectrum).

5



RF2424

5

MODULATORS AND UPCONVERTERS

Evaluation Board Layout 2" x 2"



RF2424